Equalization Techniques for MIMO Systems in Wireless Communication: A Review

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Abstract—The main objective of this paper is to provide a review of equalization techniques in multiple input, multiple output (MIMO) communication system. In wireless communication inter symbol interference is major obstacle which greatly affect the data quality. The chief goal of equalization techniques is to rebuild the actual signal with the help of filter or any other methods and remove the effect of ISI so that the reliability of data transmission is maintained. Different kind of equalization techniques proposed earlier has been reviewed here.

Keywords- MIMO, interference, MMSE, Adaptive, filter, ISI

I. INTRODUCTION

Main aim of wireless technology is to offer better quality voice, data, pictures, fax, and video. Terrestrial digital T.V. broadcasting, LTE, 3GPP are some of the latest advancement in the field of wireless communication. These above mentioned advancement are possible with the help of OFDM and CDMA technology. In a last few years OFDM has emerged as a most popular technique of transmitting data/signal over wireless media. In OFDM, signal are transmitted in sub-channel of different frequency in parallel fashion. The frequency of sub-channel are so selected that these frequency are orthogonal to each other and therefore do not interfere with each other. This phenomenon makes it possible to transmit the data in overlapping frequency and hence reduced the bandwidth requirement considerably [1]. Since in wireless communication data are sent in radio space, the channel exhibits multipath fading phenomenon which produces inter-symbol interference (ISI) in the signal received at the receiver side. Inter-symbol interference (ISI) is undesirable and it increases the bit error rate. Whenever the modulation bandwidth exceeds the radio channel coherence bandwidth, ISI produced. In order to reduce or eliminate inter-symbol interference (ISI), different types of equalization techniques are used which compensate the ISI using impulse response of channel [2]. Equalizer work by keeping bit error rate as low as possible and SNR as high as possible[3]. These equalization techniques proved to be very important for designing wireless system with high data rate transmission capacity. Most of the wireless receiver are equipped with the equalizer which gives good result under expected. In wireless communication, the receivers are generally equipped with the equalizer which gives average performance. The quality of wireless communication depends upon the three parameters i.e. rate, range and reliability of transmission. These parameter are related with each other. Simultaneously improvement in all the three parameters can be accomplished with the help of new technique called MIMO assisted OFDM.

II. BACKGROUND

The wireless communication concept came into existence when In 1928 Nyquist laid down the theory of telegraph transmission [8] which also became the foundation of pulse transmission. Widrow and Hoff [9] proposed least mean square based adaptive filter algorithm in 1960 which was the foundation of adaptive equalization method. Most of the research on adaptive equalization in early 60’s were concentrated on the basic fundamental and structure of zero forcing transversal equalizer[10]. Meanwhile in parallel some work has been done in the same period for developing the theory and structure of linear transmit and receive filters[11][12] which can minimize the mean square error for additive Gaussian noise channels[13]. In early 70’s Transversal and decision-feedback equalizing process with forward filter tap spacing which is less than the interval of symbol has been proposed[14]. These equalizers were later on used in commercial telephone line modem[15][16] and in military radio system[17]. Until the early part of 70’s most of the work related with equalization were centered to the...
structure and steady state analysis. After that most of the work were centered on developing some computationally efficient algorithm of equalization. RLS algorithm was one of the computationally efficient method of equalization which became the intense research subject[18][19][20][21][22] in the field of equalization at that time which eventually led to the transversal [23][24][25] and lattice[26][27][28][29][30][31] algorithm. Later these method were applied to adaptive equalizer for HF modem[32]. An equalization operation is generally performed at baseband or IF in a receiver section. A lot of work has been proposed in the past for combating and reducing inter symbol interference in MIMO communication system. Equalization is one of them. Equalization techniques is divided into three part. The first one frequency domain equalization techniques which includes pre-tone equalization and it is used where channel order is larger than the cyclic prefix. The second one is time domain equalization techniques which includes channel shortening and time domain statistics-based techniques. The third one is turbo equalization which is iterative based equalization technique. Some of the noteworthy algorithm of equalization is presented in the next sections of this paper.

III. ADAPTIVE EQUALIZER [33]
Randomness and time varying property of fading channel makes it necessary to design a equalizer which is capable of tracking the time varying property of mobile channel and hence known as adaptive equalizers. Block diagram of adaptive equalization process[33] is shown in Figure 1.2.

![Figure 1.2 Adaptive Equalizer](image)

If s(t) is the original signal and f(t) represent the collective complex base-band impulse response of channel, Transmitter and IF/RF section of receiver, then the signal received at the input of equalizer is given by

\[ y(t) = s(t) * f(t) + n(t) \]

Here \( f(t) \) is complex conjugate of f(t) and \( n(t) \) is noise of baseband while symbol \( * \) represents the convolution operation. The output of an equalizer can be represented by

\[ y(t) = x(t) * h_{eq}(t) \]

or it can be written as

\[ y(t) = s(t) * f(t) * h_{eq}(t) + n(t) * h_{eq}(t) \]

where \( h_{eq}(t) \) represent the equalizer impulse response. If the combined response of transmitter, communication channel and IF/RF section of receiver is represented by \( g(t) \) then the above equation can also be written as

\[ y(t) = h(t) * g(t) + n(t) * h_{eq}(t) \]

Impulse response of the equalizers transversal filter can be expressed as

\[ h_{eq}(t) = \sum c_n \delta(t - nT) \]

Here \( c_n \) represent the complex coefficients of filter at the equalizer.

Since the desired output signal of the equalizer is nothing but the original signal s(t). Therefore in order to make \( y(t) = s(t) \) in equation

\[ y(t) = s(t) * f(t) * h_{eq}(t) + n(t) * h_{eq}(t) \]

the value of g(t) must be

\[ g(t) = f(t) * h_{eq}(t) \delta(t) \]

the main aim of equalizer is to fulfill the following equation

\[ H_{eq}(f) F(-f) = 1 \]

Where \( H_{eq}(f) \) and \( F(f) \) represent Fourier transforms of \( h_{eq}(t) \) and \( f(t) \) respectively.

IV. ZERO FORCING EQUALIZER [34]
In this method [34], the ISI component at the output of equalizer is forced to zero by using appropriate linear time invariant filter having suitable transfer function. If transmitted symbol is represented by \( x_1 \) and \( x_2 \), \( h_{11} \) represent the channel from first transmitter to first receiver, \( h_{12} \) represent the channel from second transmitter to first receiver, \( h_{21} \) represent the channel from first transmitter to second receiver and \( h_{22} \) represent the channel from second transmitter to second receiver and \( n_1, n_2 \) represent noise on first and second receiver then the received signal on first receiver is given by

\[ y_1 = h_{11} x_1 + h_{12} x_2 + n_1 \]

\[ = [h_{11} \; h_{12}] [x_1 \; x_2] + n_1 \]

And the received symbol on second receiver is given by

\[ y_2 = h_{21} x_1 + h_{22} x_2 + n_2 \]

\[ = [h_{21} \; h_{22}] [x_1 \; x_2] + n_2 \]

These two above equation can also be written as

\[ \begin{bmatrix} y_1 \\ y_2 \end{bmatrix} = \begin{bmatrix} h_{11} & h_{12} \\ h_{21} & h_{22} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} + \begin{bmatrix} n_1 \\ n_2 \end{bmatrix} \]

It is clear from this equation that if \( h_{11}, h_{12}, h_{21}, h_{22} \) and \( y_1, y_2 \) is known then it is easier for the receiver to compute the \( x_1 \) and \( x_2 \).

Now if we rewrite the above equation then

\[ y = Hx + n \]

From here it is clear that in order to find \( x \) from above equation, we need to find out the matrix which is inverse of matrix H. If W represent the inverse of H then it must satisfy the property

\[ WH=W \]

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Where the matrix \( W \) is the identity matrix. The matrix \( W \) which satisfies the above mentioned property is known as the zero forcing linear detector and is computed by following equation
\[
W = (H^H H)^{-1} H^H
\]
In this equation the matrix \( H^H H \) is given by
\[
H^H H = [h_{11}^* h_{12}^* h_{11} h_{12}]
\]
From this matrix it is clear that the off diagonal term are non zero and hence zero forcing equalizer cancel out the interference signal. It is reasonably simple and easy to implement but its main drawback is that it tends to amplify the noise and hence gives noisy output.

V. MMSE EQUALIZER (MINIMUM MEAN SQUARED ERROR)

This type of equalizer uses the squared error as performance measurement\[35\]. The receiver filter is designed to fulfill the minimum mean square error criterion. Main objective of this method is to minimize the error between target signal and output obtained by filter. The computation for this method is as follows- If transmitted symbol is represented by \( x_1 \) and \( x_2 \), \( h_{11} \) represent the channel from first transmitter to first receiver, \( h_{12} \) represent the channel from second transmitter to first receiver, \( h_{21} \) represent the channel from first transmitter to second receiver and \( h_{22} \) represent the channel from second transmitter to second receiver and \( n_1, n_2 \) represent noise on first and second receiver then the received symbol on first receiver is given by
\[
y_1 = h_{11} x_1 + h_{12} x_2 + n_1
\]
\[
= [h_{11} h_{12}] [x_1 x_2] + n_1
\]
And the received symbol on second receiver is given by
\[
y_2 = h_{21} x_1 + h_{22} x_2 + n_2
\]
\[
= [h_{21} h_{22}] [x_1 x_2] + n_2
\]
These two above equation can also be written as
\[
[y_1 y_2] = [h_{11} h_{12}] [x_1 x_2] + [n_1 n_2]
\]
It is clear from this equation that if \( h_{11}, h_{12}, h_{21}, h_{22} \) and \( y_1, y_2 \) is known then it is easier for the receiver to compute the \( x_1 \) and \( x_2 \).

Now if we rewrite the above equation then
\[
y = H x + n
\]
Now, MMSE algorithm computes the coefficient of matrix \( W \) which minimize the condition
\[
E \left[ \| w_y - x \| \| w_y - x \|^H \right]
\]
Solving above equation gives
\[
W = (H^H H + N_o I)^{-1} H^H
\]
From the above equation it is clear that this equation is different from the equation of zero forcing equalizer by the term \( N_o I \). If we put \( N_o I = 0 \) in this equation then MMSE equalizer becomes zero forcing equalizer.

VI. ZERO FORCING WITH SUCCESSIVE INTERFERENCE CANCELLATION (ZF-SIC)

In this method, first of all the zero forcing equalizer find the estimated symbol \( \hat{x}_1 \) and \( \hat{x}_2 \) then one of the estimated symbol is subtracted from received symbol to compute the equalized symbol by applying maximum ration combining(MRC)\[36\].

If \( x_1 \) and \( x_2 \) are the estimated transmitted symbol then
\[
[\hat{\gamma}_1 \hat{\gamma}_2] = (H^H H)^{-1} H^H [\gamma_1 \gamma_2]
\]
By subtracting one of the estimated symbol (say \( \hat{x}_1 \)) from the received signal \( y_1 \) and \( y_2 \)
\[
\begin{align*}

r_1 &= y_1 - h_{12} \hat{x}_2 \\

r_2 &= y_2 - h_{22} \hat{x}_2
\end{align*}
\]
\[
\begin{align*}

[\tilde{r}_1 \tilde{r}_2] &= [h_{12} x_1 + n_1] \\

[\tilde{r}_1 \tilde{r}_2] &= [h_{21} x_1 + n_2]
\end{align*}
\]
Or
\[
r = h x_1 + n
\]
By applying maximum ratio combining(MRC), the equalized symbol is given by
\[
\hat{x}_1 = \frac{h^H r}{h^H h}
\]

VII. SUCCESSIVE INTERFERENCE CANCELLATION USING OPTIMAL ORDERING [37]

In the previous successive interference cancellation method, estimation symbol is chosen arbitrarily and then its effect is subtracted from received symbol \( y_1 \) and \( y_2 \). A better result can be obtained if we choose estimated symbol whose influence is more than other symbol. For this first of all the power of both the symbol is computed at the receivers and then the symbol having higher power is chosen for subtraction process.

The power of transmitted symbol \( x_1 \) is given by
\[
P_{x_1} = h_{11}^2 + h_{21}^2
\]
Similarly the power of transmitted symbol \( x_2 \) is given by
\[
P_{x_2} = h_{12}^2 + h_{22}^2
\]
If \( P_{x_1} > P_{x_2} \), the \( x_1 \) is subtracted from \( y_1 \) and \( y_2 \) and re-estimate the \( \hat{x}_2 \) i.e.
\[
\begin{align*}

[\tilde{r}_1 \tilde{r}_2] &= [y_1 - h_{11} \hat{x}_1] \\

[\tilde{r}_1 \tilde{r}_2] &= [y_2 - h_{12} \hat{x}_1]
\end{align*}
\]
\[
\begin{align*}

[\tilde{r}_1 \tilde{r}_2] &= [h_{21} \hat{x}_1 + n_1] \\

[\tilde{r}_1 \tilde{r}_2] &= [h_{22} \hat{x}_1 + n_2]
\end{align*}
\]
Or
\[
r = h x_2 + n
\]
By applying maximum ratio combining (MRC), the equalized symbol is given by
\[
\hat{x}_2 = \frac{h^H r}{h^H h}
\]
Similarly if \( P_{x_2} > P_{x_1} \), the \( x_2 \) is subtracted from \( y_1 \) and \( y_2 \) and re-estimate the \( \hat{x}_1 \). These are some of the most
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